

Variable speed wind turbine control scheme using a robust wind torque estimation

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Abstract

1 This work proposes a robust controller for a variable speed wind turbine system
2 with a doubly fed induction generator. The controller aims at tracking the optimal
3 speed of the wind turbine so that extracts the maximum power from the wind. Also,
4 a robust aerodynamic torque observer is proposed in order to avoid the use of wind
5 speed sensors. This torque observer allows to estimate the aerodynamic torque to be
6 used by the controller in order to calculate the value of the optimal reference speed
7 for the wind turbine. The vector control theory is applied in the present approach, and
8 thereby the stator flux-oriented control is used for controlling the speed of the wind
9 turbine generator. The proposed robust control law is based on sliding mode control
10 theory, which has proved to provide good performance under system uncertainties.

11 The stability of the proposed controller under disturbances and parameter uncer-
12 tainties has been analyzed using the Lyapunov stability theory. Finally, real time ex-
13 perimental results show that, on the one hand, the proposed controller provides high-
14 performance dynamic characteristics, and on the other hand, this scheme is robust with
15 respect to the uncertainties that usually appear in this kind of systems.

16 **Keywords:** Renewable Energy; Wind Energy; Sliding Mode Observer; Sliding
17 Mode Control; DFIG; Lyapunov Stability.

1. Introduction

19 There have been amazing improvements in wind energy extraction during the last
20 decades. There are three main factors that explain this development. First, the decline
21 of fossil fuel reserves and the pollution they generate. This scenario stimulated the
22 research in alternative sources of energy like photovoltaic energy and wind energy.
23 Second, wind energy can be found anywhere on the Earth and, in some locations,
24 presenting a considerable density of energy. Third, some governments promoted its
25 development for geostrategic reasons.

26 As a result, the wind power capacity installed has been increased intensively through-
27 out the last decades. As a matter of example, the worldwide installed wind power ca-
28 pacity was about 7480 MW in 1997 whereas in the year 2017 there was around 487
29 GW. During the first years of the period from 1997 to 2016 the wind capacity was in-
30 creased at the rate of 2000 MW per year. However, the annual growth rate increased
31 significantly to reach a top of 64 GW, in year 2015, keeping slightly lower during the
32 year 2016 at 55 GW. The increment of year 2016 amounts around 12% of the total
33 wind power capacity installed worldwide and is the second largest to date just be-
34 low the maximum of 2015. A significant decline in the Chinese market, following a
35 very strong 2015, was responsible for most of the market contraction. Even so, China
36 retained its lead for new installations, followed distantly by the United States and Ger-
37 many, with India passing Brazil. Others in the top 10 for additions were France, Turkey,
38 the Netherlands, the United Kingdom and Canada.

39 In 2016, China (ranked first country for additions) added 23.4 GW of wind energy,
40 approaching its total installed capacity to 169 GW. China accounted for one-third of
41 the total global capacity by the end of this year. However, new installations were
42 down 24% relative to 2015, when the top growth rate was observed. The United States,
43 ranked second for additions (8.2 GW), had a cumulative capacity of 82.1 GW at the end
44 of 2016. Regarding energy generation the United States produced 226.5 TWh which
45 was only 6% below China, during 2016. Germany again was the largest European
46 market, and the third in the world, increasing operating wind power capacity by almost
47 5 GW for a total of 49.5 GW [1].

48 Therefore, wind energy for electric power generation is an area of research interest.
49 Nowadays, the emphasis is given to the cost-effective utilization of this energy aiming
50 at ensuring quality and reliability in the electricity delivery. Variable speed wind tur-
51 bines (VSWT) are continuously increasing their market share. This kind of turbines
52 allow to track the changes in the speed of the wind by adapting the shaft speed and
53 thus maintaining the optimal power generation. However, a VSWT needs an adequate
54 control system in order to operate at the optimal power efficiency [2], [3], [4], [5], [6].

55 The control systems for variable speed wind turbines may benefit from Doubly
56 Feed Induction Generators (DFIG), since they allow maximizing wind power extrac-
57 tion. Vector control techniques are frequently adopted for that purpose. This kind of
58 controllers ease achieving these objectives: (1) regulating the rotor speed in order to
59 maximize the extracted power, (2) keeping constant the frequency of the output voltage
60 at the DFIG stator, and (3) governing the reactive power extracted from the DFIG [7],
61 [8], [9], [10].

62 The controller is responsible for carrying the turbine rotor speed into the desired
63 optimal speed that maximizes the active power extracted from the wind, despite system
64 uncertainties and wind velocity variations [11], [12], [13]. This paper explores a new
65 approach for variable speed wind turbines based on a robust speed control method.
66 The proposed robust control is based on sliding mode control theory, which has proved
67 to present a good behaviour for systems subjected to uncertainties. Moreover, this
68 kind of controllers are adequate for real time applications when limited computational
69 resources are available, since it presents low computational costs [14], [15].

70 In order to calculate the optimal rotor speed, the wind speed value should be mea-

71 sured or estimated [16], [17], [18], [19]. This is precisely one of the contributions of
72 this work when compared with our previous works [20] [21]. In this new paper an aero-
73 dynamic torque observer is proposed so that the control relies on the estimated aero-
74 dynamic torque rather than the wind speed measurements. In our previous research,
75 sensors were always used to obtain the wind speed in order to calculate the reference
76 speed for the turbine. Also, this work validates experimentally the combination of the
77 robust control scheme for wind speed control with the proposed torque observer. For
78 that purpose, different experiments were developed over a test bench specifically de-
79 signed and constructed. The analysis of the experimental results, obtained at several
80 operational conditions, guarantees a satisfactory behaviour.

81 2. System modelling

82 The capacity of a wind turbine for extracting power from the wind depends of three
83 major factors: (1) the available power of the wind, (2) the power curve of the wind
84 turbine, and (3) the ability to respond to wind fluctuations. The mechanical power
85 produced by the wind is given by the following expression [22], [23]:

$$P_m(v, \lambda, \beta) = \frac{1}{2} C_p(\lambda, \beta) \rho \pi R^2 v^3 \quad (1)$$

86 where the radius of the rotor is R , the density of the air is ρ , the wind velocity is
87 v , the wind turbine power coefficient is C_p , the pitch angle is β and, finally, λ is the
88 tip-speed ratio, given by the following expression:

$$\lambda = \frac{R w}{v} \quad (2)$$

89 in which w represents the speed of the turbine rotor. This expression shows that,
90 assuming a constant value for the rotor speed, the variation at the wind speed produce
91 changes in the tip-speed ratio, since it modifies the power coefficient C_p , and, conse-
92 quently, the power extracted from the wind turbine. Thus, the optimal tip-speed ratio,
93 which extracts the maximum power output, could be kept constant by maintaining the
94 ratio between the rotor speed and the wind speed.

95 The torque generated by the wind turbine can be obtained by combining expres-
96 sions (1) and (2) into:

$$T_m(v, \lambda, \beta) = \frac{P_m(v, \lambda, \beta)}{w} = \frac{1}{2} C_p(\lambda, \beta) \rho \pi \frac{R^3}{\lambda} v^2 \quad (3)$$

97 In a simplified way, typical wind power generation systems are composed by the
98 following three elements: (1) turbine, which converts wind energy into mechanical
99 energy, (2) gearbox, which increases the speed and decreases the torque, and (3) gen-
100 erator, which produces electrical energy from mechanical energy.

101 The input wind torque T_m drives the wind turbine rotor to spin at the speed w ,
102 producing a transmission output torque T_t which is used to fed the generator. The shaft
103 torque at the generator T_e produces an angular velocity of w_e . Note that, the rotor and
104 generator speeds may differ since they are related by a gearbox.

105 The mechanical behaviour of the wind power system may be characterized by the
 106 following equations[24]:

$$J_m \dot{w} + B_m w = T_m - T \quad (4)$$

$$J_e \dot{w}_e + B_e w_e = T_t - T_e \quad (5)$$

$$T_t w_e = T w \quad (6)$$

107 where J_e and J_m are the moments of inertia of both generator and turbine, B_e and
 108 B_m represent respectively the viscous friction coefficients of generator and turbine,
 109 T_m is the torque generated by the wind at the turbine, the torque at the transmission
 110 shaft before and after the gear box are represent by T and T_f respectively and T_e is the
 111 torque produced by the generator when spinning at angular velocity w_e .

112 The angular velocities of both generator, w_e , and turbine, w , are related by the gear
 113 ratio η :

$$\gamma = \frac{w_e}{w} \quad (7)$$

114 The following expression, which represents the model of the wind system, can be
 115 obtained by combining equations (4), (5), (6) and (7):

$$J \dot{w} + B w = T_m - \gamma T_e \quad (8)$$

116 where

$$J = J_m + \gamma^2 J_e \quad (9)$$

$$B = B_m + \gamma^2 B_e \quad (10)$$

117 3. Aerodynamic torque observer

118 This section proposes an aerodynamic torque estimator aimed at avoiding the wind
 119 speed measurements needed to calculate the optimal turbine speed that extracts the
 120 maximum power from the wind.

121 The aerodynamic torque may be considered as a quasi-constant signal for a time
 122 interval since the variation of the mean value of the wind does not change quickly and,
 123 moreover, the wind turbine system has some inertia that may absorb these variations.
 124 Accordingly, the state space equations for the wind turbine system (8) may be rewritten
 125 as:

$$\begin{aligned} \dot{w} &= \frac{1}{J} (T_m - B w - \gamma T_e) \\ \dot{T}_m &= 0 \end{aligned} \quad (11)$$

126 Since the aerodynamic torque T_m is taken as a quasi-constant signal, it can be regarded
 127 as the slow component of the system. According to singular perturbation theory [25],
 128 the stability of a system can be proved if the asymptotic stability of the fast component
 129 (i.e. rotor speed) is ensured. Thereafter, for the reduced system, the convergence of the

130 slow component (i.e. aerodynamic torque) can be achieved when the estimation error
 131 for the rotor speed is null.

132 The next observer, based on sliding mode theory, is proposed:

$$\begin{aligned}\dot{\hat{w}} &= \frac{1}{J} \left(\hat{T}_m - Bw - \gamma T_e \right) + k_{w_1} e_w + h_1 \operatorname{sgn}(e_w) \\ \dot{\hat{T}}_m &= k_{w_2} e_w + h_2 \operatorname{sgn}(e_w)\end{aligned}\quad (12)$$

133 where the observation error of the turbine speed is $e_w = w - \hat{w}$, and $h_1, h_2, k_{w_1}, k_{w_2}$
 134 are the parameters of the observer, which should be chosen greater than 0.

135 The observation error dynamics for the rotor speed may be obtained by subtracting
 136 eqn.(12) from eqn.(11):

$$\begin{aligned}\dot{e}_w &= \frac{1}{J} e_T - k_{w_1} e_w - h_1 \operatorname{sgn}(e_w) \\ \dot{e}_T &= -k_{w_2} e_w - h_2 \operatorname{sgn}(e_w)\end{aligned}\quad (13)$$

137 where $e_T = T_m - \hat{T}_m$

138 The stability of the fast component can be proved by means of the following Lya-
 139 punov candidate function.

$$V = \frac{1}{2} e_w^2 \quad (14)$$

140 The derivative of this function with respect to time is:

$$\dot{V} = e_w \dot{e}_w \quad (15)$$

$$= e_w \left(\frac{1}{J} e_T - k_{w_1} e_w - h_1 \operatorname{sgn}(e_w) \right) \quad (16)$$

$$= \frac{1}{J} e_w e_T - h_1 |e_w| - k_{w_1} e_w^2 \quad (17)$$

141 The next condition must be satisfied to ensure that \dot{V} is a negative definite function:

$$h_1 \geq \left| \frac{1}{J} e_T \right| - k_{w_1} |e_w| + \eta_w, \quad \eta_w > 0 \quad (18)$$

142 Accordingly,

$$\dot{V} \leq -\eta_w |e_w| \quad (19)$$

143 From (19) it may be deduced that the equilibrium point, $e_w = 0$, is asymptotically
 144 stable. Also, this equation shows that the equilibrium point $e_w = 0$ is reached in finite
 145 time since:

$$t_{reach} \leq \frac{e_w(t=0)}{\eta_w} \quad (20)$$

146 From (13) it may be noticed that when the equilibrium point is reached, i.e. $e_w = 0$
 147 and $\dot{e}_w = 0$, the dynamics of the error observer is comparable to the following reduced-
 148 order subsystem:

$$0 = \frac{1}{J}e_L - h_1 \operatorname{sgn}(e_w) \quad (21)$$

$$\dot{e}_T = -h_2 \operatorname{sgn}(e_w) \quad (22)$$

149 Using the above equations it can be inferred:

$$\dot{e}_T = \frac{-1}{J} \frac{h_2}{h_1} e_T \quad (23)$$

150 In conclusion, the aerodynamic torque estimation error converges to zero if the ob-
 151 server gains h_1 , h_2 k_{w_1} and k_{w_2} are appropriately selected. Accordingly, the estimated
 152 states \hat{w}_m , \hat{T}_m converge to the real states w_m , T_L as t tends to infinity. So, the aero-
 153 dynamic torque may be obtained from the sliding mode observer given by eqn.(12).

154 4. Wind turbine generator control

155 In this VSWT system a DFIG is employed, which is fed from the rotor and the stator
 156 of the generator. The rotor is fed by means of a variable frequency converter (VFC)
 157 and the stator is directly connected to the grid. This system is required to deliver
 158 electrical power to the grid over a wide range of operation (from subsynchronous to
 159 supersynchronous speed) while ensuring constant values for voltage and frequency.
 160 For that purpose, the active power flow between the rotor circuit and the grid should
 161 be controlled in magnitude and direction. This is achieved by two four-quadrant IGBT
 162 PWM converters, one for the rotor-side (RSC), and another for the grid-side (GSC),
 163 connected back-to-back by a DC-link capacitor [26].

164 In order to extract the maximum power from the wind, improving the efficiency, the
 165 shaft speed of the generator must be regulated in order to obtain the optimal tip-speed
 166 ratio λ_{opt} . This optimal tip-speed ratio maximizes the value of the power coefficient
 167 $C_{p_{max}}$ and, therefore, the generated power[13]. In other words, there is a unique value
 168 for the wind turbine speed that, for a particular wind speed, maximizes wind power
 169 extraction. The maximum value of the power coefficient curves versus tip-speed ratio
 170 is represented by λ_{opt} . This value depends of the turbine design characteristics.

171 In this scenario, the use of the estimated aerodynamic torque \hat{T}_m allows calculating
 172 the turbine speed reference without using wind speed measurements.

173 Equations (1) and (2) may be combined to express the wind turbine power as a
 174 function of the turbine speed:

$$P_m(w, \lambda, \beta) = \frac{1}{2} C_p(\lambda, \beta) \rho \pi R^2 \left(\frac{Rw}{\lambda} \right)^3 = k_w(\lambda, \beta) \cdot w^3 \quad (24)$$

where

$$k_w = \frac{1}{2} C_p(\lambda, \beta) \rho \pi \frac{R^5}{\lambda^3}$$

175 Thus, the value for the maximum power generation can be calculated with eqn.(24):

$$P_{m_{opt}} = T_{m_{opt}} \cdot w_{opt} = k_{w_{opt}} \cdot w_{opt}^3 \quad (25)$$

where

$$k_{w_{opt}} = \frac{1}{2} C_p(\lambda_{opt}, \beta) \rho \pi \frac{R^5}{\lambda_{opt}^3}$$

176 being λ_{opt} the value of λ that yields the maximum power coefficient C_p .

177 Hence, the optimal speed for the rotor can be obtained from eqn.(25) and the estimated aerodynamic torque (12):
178

$$w_{opt} = \sqrt{\frac{\hat{T}_m}{k_{w_{opt}}}} \quad (26)$$

179 Figure 1 shows the power characteristic curves for the turbine employed in this
180 study, namely, several curves for wind speed values ranging from 5 m/s to 16.2 m/s. It
181 should be noted that eqn.(26) yields the points of the red curve of figure 1 between B
182 and C since they are the maximum values of the turbine output power.

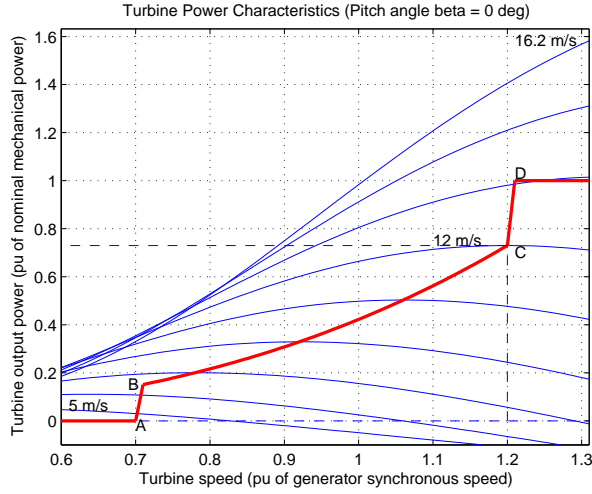


Figure 1: Turbine Power Characteristics

183 Therefore, a control scheme aimed at regulating the wind turbine speed in order to
184 track the optimal wind speed is required. This controller should be able to maximize
185 the power coefficient and, therefore, to extract the maximum power from the wind,
186 increasing the wind turbine performance. Consequently, the turbine speed controller
187 should be adequately designed in order to track the optimal reference speed for the
188 turbine $w^* = w_{opt}$.

189 For this purpose, the current of the rotor is typically regulated using the stator-
190 flux oriented reference frame [27]. In this frame, the d-axis is aligned with the stator

191 flux linkage vector ψ_s , so that $\psi_{ds}=\psi_s$ and $\psi_{qs}=0$ and accordingly the next model is
 192 obtained [28]:

$$i_{qs} = \frac{L_m i_{qr}}{L_s} \quad (27)$$

$$i_{ds} = \frac{L_m (i_{ms} - i_{dr})}{L_s} \quad (28)$$

$$T_e = \frac{L_m i_{ms} i_{qr}}{L_s} \quad (29)$$

$$Q_s = \frac{3 w_s L_m^2 i_{ms} (i_{ms} - i_{dr})}{2 L_s} \quad (30)$$

$$v_{dr} = r_r i_{dr} + \sigma L_r \frac{di_{qr}}{dt} - s w_s \sigma L_r i_{qr} \quad (31)$$

$$v_{qr} = r_r i_{qr} + \sigma L_r \frac{di_{qr}}{dt} \quad (32)$$

$$+ s w_s \left(\frac{\sigma L_r i_{dr} + L_m^2 i_{ms}}{L_s} \right) \quad (33)$$

193 where

$$i_{ms} = \frac{v_{qs} - r_s i_{qs}}{w_s L_m} \quad (34)$$

$$\sigma = 1 - \frac{L_m^2}{L_s L_r} \quad (35)$$

194 Given that the grid is connected to the stator and considering that the the stator resis-
 195 tance has low influence, the value of the magnetizing current of the stator i_{ms} can be
 196 approximated as a constant value [26]. Accordingly, the electromagnetic torque of the
 197 DFIG can be modeled as:

$$T_e = K_T i_{qr} \quad (36)$$

198 where K_T is a torque constant defined as:

$$K_T = \frac{L_m i_{ms}}{L_s} \quad (37)$$

199 From equations (8) and (36) it may be observed that the q-axis rotor current component
 200 i_{qr} can be used to control the speed of the wind turbine. Equation (30) shows that the
 201 d-axis component of the rotor current i_{dr} could be used to govern the reactive power at
 202 the stator Q_s . Accordingly, the values for the currents i_{qr} and i_{dr} can be obtained from
 203 the reference values provided from w_r and Q_s .

204 The following differential equation for the system speed has been inferred from
 205 equations (8) and (36):

$$\dot{w} = \frac{1}{J} (T_m - \gamma K_T i_{qr} - Bw) \quad (38)$$

$$= -aw + f - bi_{qr} \quad (39)$$

206 being the parameters f , a and b defined as:

$$f = \frac{T_m}{J}, \quad a = \frac{B}{J}, \quad b = \frac{\gamma K_T}{J}; \quad (40)$$

207 Now, the previous equation (39) will be considered taking into account some un-
208 certainties:

$$\dot{w} = -(a + \Delta a)w + (f + \Delta f) - (b + \Delta b)i_{qr} \quad (41)$$

209 where the terms Δb , Δa and Δf take into account the uncertainties at the parameters
210 b , a and f respectively.

211 Let be defined the speed tracking error as:

$$e(t) = w(t) - w^*(t) \quad (42)$$

212 where the command for the rotor speed is w^* .

213 Calculating the time derivative of the previous equation it is obtained:

$$\dot{e}(t) = \dot{w} - \dot{w}^* = -a e(t) + u(t) + d(t) \quad (43)$$

214 where the signal $u(t)$ groups the know terms,

$$u(t) = f(t) - b i_{qr}(t) - a w^*(t) - \dot{w}^*(t) \quad (44)$$

215 and the signal $d(t)$ groups the uncertainty terms,

$$d(t) = -\Delta a w(t) + \Delta f(t) - \Delta b i_{qr}(t) \quad (45)$$

216 The system uncertainties, described above, may be compensated by a sliding control
217 scheme. Note that in sliding mode control theory the sliding gain must be selected in
218 order to satisfy the sliding condition [29]. Hence, the value for the sliding gain should
219 be chosen carefully to meet the previous condition.

220 The proposed sliding variable $S(t)$ is defined with an integral component as:

$$S(t) = e(t) + \int_0^t (k + a)e(\tau) d\tau \quad (46)$$

221 where k is a positive constant gain.

222 Regarding the sliding surface, it has been defined as:

$$S(t) = e(t) + \int_0^t (a + k)e(\tau) d\tau = 0 \quad (47)$$

223 Also, a variable structure speed controller, aimed at controlling the wind turbine speed,
224 has been introduced.

$$u(t) = -k e(t) - \beta \operatorname{sgn}(S) \quad (48)$$

225 where β is the sliding gain and $\operatorname{sgn}(\cdot)$ is the sign function.

226 The next assumption should be satisfied to obtain the tracking of the optimal refer-
227 ence speed for the wind turbine:

228 (A1) The gain β must be selected so that $\beta \geq |d(t)|$.

229 Note that this assumption means that the value of the system uncertainties are
230 finite.

231 **Theorem 1.** *Considering a VSWT defined by equation (41) and assumption (A1).
232 Then, the control law presented in Eqn.(48) governs the speed of the wind turbine
233 $w(t)$ so that the tracking error of the wind turbine velocity $e(t) = w(t) - w^*(t)$ ap-
234 proximates to zero exponentially.*

235 This theorem may be proved by means of the Lyapunov stability theory.

236 **Proof :** The candidate of the Lyapunov function is defined as:

$$V(t) = \frac{1}{2}S(t)S(t) \quad (49)$$

237 Calculating the time derivative of the previous function:

$$\begin{aligned} \dot{V}(t) &= S(t)\dot{S}(t) \\ &= S \cdot [\dot{e} + (k + a)e] \\ &= S \cdot [(-a e + u + d) + (k e + a e)] \\ &= S \cdot [u + d + k e] \\ &= S \cdot [-k e - \beta \operatorname{sgn}(S) + d + k e] \\ &= S \cdot [d - \beta \operatorname{sgn}(S)] \\ &\leq -(\beta - |d|)|S| \\ &\leq 0 \end{aligned} \quad (50)$$

238 Eqns. (46), (43) and the (A1) assumption were used in this proof.

239 Previous equations show that (1) $V(t)$ is positive-definite, (2) $\dot{V}(t)$ is negative def-
240 inite and (3) $V(t)$ tends to infinity when $S(t)$ tends to infinity. From the Lyapunov
241 direct method, it may be concluded that the equilibrium at the origin $S(t) = 0$ is
242 globally asymptotically stable. Hence, when time tends to infinity, the sliding variable
243 $S(t)$ goes to zero. Besides, the trajectories of this system will reach the sliding surface
244 $S = 0$ in finite time and thereafter will remains in this surface (47). When so happens
245 $S(t) = \dot{S}(t) = 0$ and therefore the dynamic behaviour of the tracking problem (43) is
246 represented by the next dynamic equation:

$$\dot{S}(t) = 0 \quad \Rightarrow \quad \dot{e}(t) = -(k + a)e(t) \quad (51)$$

247 Therefore, taking into account that k and a are a positive constants, the speed track-
248 ing error $e(t)$ converges exponentially to zero.

249 Finally, the command for q-component of the current, $i_{qr}^*(t)$, can be calculated
250 from equations (48) and (44):

$$i_{qr}^*(t) = \frac{1}{b} [k e + \beta \operatorname{sgn}(S) - a w^* - \dot{w}^* + f] \quad (52)$$

251 Therefore, the wind turbine speed regulation in order to maximize the power gen-
 252 eration under system uncertainties can be obtained using the value for the rotor current
 253 given by the previous equation.

254 5. Experimental Results

255 This section analyses the behaviour of the proposed control scheme when the slid-
 256 ing mode observer developed in this paper is used, by means of several experimental
 257 tests in a real platform designed ad hoc.

258 5.1. Description of the platform

259 Figure 2 shows the block diagram of the experimental platform used for the valida-
 260 tion of the control set up. A photography of the experimental platform is provided at
 261 figure 3.

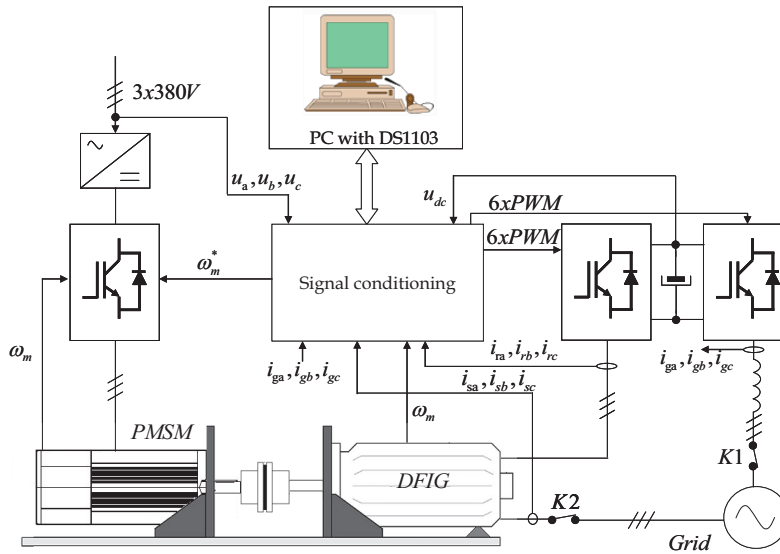


Figure 2: Block diagram of the experimental platform

262 The control platform displayed in figure 2 includes a PC with MatLab7/Simulink
 263 R2007a and DSControl 3.2.1 software and a dSpace DS1103 controller board, which
 264 includes a floating point PowerPC processor of 1 GHz. The proposed test bench also
 265 includes a commercial DFIG machine of 7.5 kW and 1447 rpm provided by Leroy

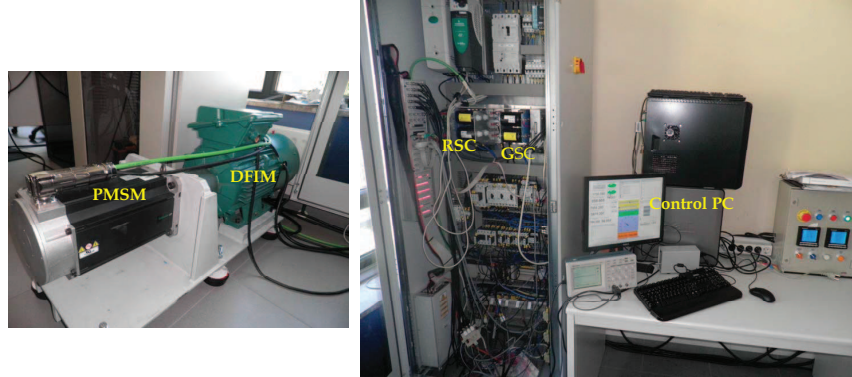


Figure 3: Photography of the experimental platform

266 Somer. This DFIG is connected to the grid through the rotor in a back-to-back con-
 267 figuration with two voltage source inverters of Dutt Power Electronics & Control. The
 268 parameters of the DFIG are indicated in Table 1.

269 The wind turbine has been emulated by means of a synchronous AC servo motor,
 270 namely a 10.6 kW 190U2 provided by Unimotor. The DFIG and the synchronous AC
 271 servo moto are mechanically coupled by the shaft.

272 The mechanical wind turbine torque values are generated using a turbine model
 273 implemented in Simulink, based on equation (3). The wind profile is used as input for
 274 the turbine model in Simulink and this model generates the torque reference for the
 275 AC servo motor. Then the AC servo motor generates the torque that moves the rotor
 276 of the DFIG. In this sense, the torque produced by a wind profile in the wind turbine is
 277 generated by means of the synchronous AC servo motor

278 A servo motor incremental encoder of 4096 square impulses per revolution is em-
 279 ployed in order to measure the rotor speed.

280 Both rotor and stator currents are limited to their nominal values to protect the
 281 machine against over currents. All sensors used to measure currents, voltages and speed
 282 magnitudes were adapted to be connected to the DS1103 controller board. This board
 283 controls both inverters generating the SVPWM (space vector pulse width modulation)
 284 pulses. The SVPWM frequency is set at 7kHz, determining the sample period for the
 285 program execution at $143 \mu s$. A dead time of $1 \mu s$, which is controlled both by software
 286 and hardware, is used for the inverters. A synchronous reference frame phase-locked
 287 loop (SRF-PLL) is used for extracting the grid voltage phase, frequency and amplitude
 288 in order to synchronize the DFIG with the grid.

289 The grid side converter is controlled in the grid voltage reference system, imposing
 290 a DC voltage of 570V and a reactive power of 0 VAR. Conventional PI controllers have
 291 been used for this task. The analysis of this control is out of scope in this work.

292 The starting process begins with the grid synchronization, when this is done the K1

293 contactor is closed and the DC voltage bus is regulated. The voltage in the DC link
 294 must be greater than the peak value of the grid voltage, which is 540 for a grid RMS
 295 voltage of 380V. Therefore, the DC voltage is regulated to 570V. Once the DC voltage
 296 is regulated, the encoder offset is determined and the DFIG stator is synchronized with
 297 the grid, connecting after that the stator to the grid closing K2. Finally, the rotor current
 298 is regulated using the implemented controllers.

Stator Voltage	380 V
Rotor Voltage	190 V
Rated stator current	18 A
Rated rotor current	24 A
Rated speed	1447 r.p.m. @ 50 Hz
Rated torque	50 Nm
Stator resistance	0.325 Ω
Rotor resistance	0.275 Ω
Magnetizing inductance	0.0664 H
Stator leakage inductance	0.00264 H
Rotor leakage inductance	0.00372 H
Inertia moment	0.07 Kg.m ²

Table 1: Ratings and parameters of the DFIG (Leroy Somer).

299 5.2. Simulation of the DFIG and validation of the real platform

300 A comparison between the response of the experimental system platform and the
 301 simulation model at different speeds and load torques proves that the model used for
 302 the experimental platform is adequate since both systems produce similar results. The
 303 test has been performed during 9 seconds when changing simultaneously the speed of
 304 the DFIG from 900 rpm to 2000 rpm and the load torque from 0 to 25 Nm.

305 Figure 4 pictures the rotor speeds, the load torque reference and the electromag-
 306 netic torque obtained from both the simulation model and the real platform. Figure
 307 5 depicts the power obtained from stator and rotor in both the simulation model and
 308 the experimental system platform for the specific speed and torque profiles. Finally,
 309 figure 6 shows the rotor d and q current components as well as the rotor phase current
 310 for the speed and load torque profiles used. The graphs obtained at simulation and in
 311 the experimental system platform look very similar, showing that the simulation model
 312 approximates accurately the real system. Therefore, the simulation model can be used
 313 to adjust the sliding controller and torque estimator, before implementing this control
 314 scheme at the real system platform avoiding undesirable damages.

315 5.3. Sliding torque estimator and optimum speed reference

316 In the real platform, the wind turbine can be modelled using equation (3), which
 317 determines the torque generated for the selected wind speed profile. The characteristic
 318 curves of the wind turbine model were presented in figure 1. Figure 7 shows the char-
 319 acteristic curves of the power coefficient $C_p(\lambda, \beta)$ for this wind turbine versus different
 320 λ and β values.

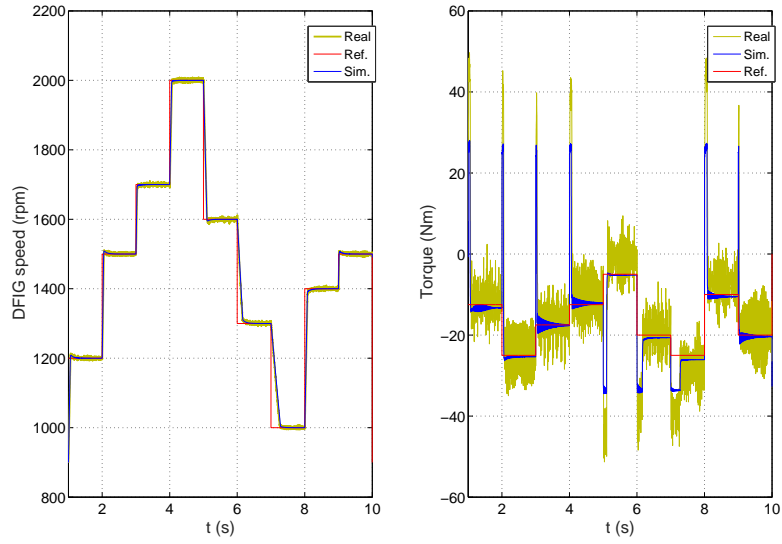


Figure 4: Real and simulated DFIG speed and torque

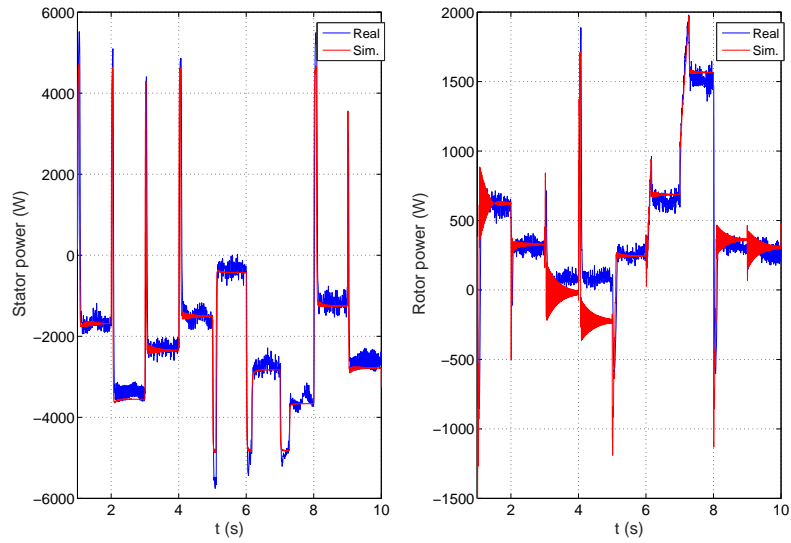


Figure 5: Real and simulated DFIG stator and rotor power for the speed and torque profile of figure 4

321 Next, the performance of the proposed sliding mode estimator is analysed. For
 322 this purpose, the wind turbine speed and torque values obtained at the experimental

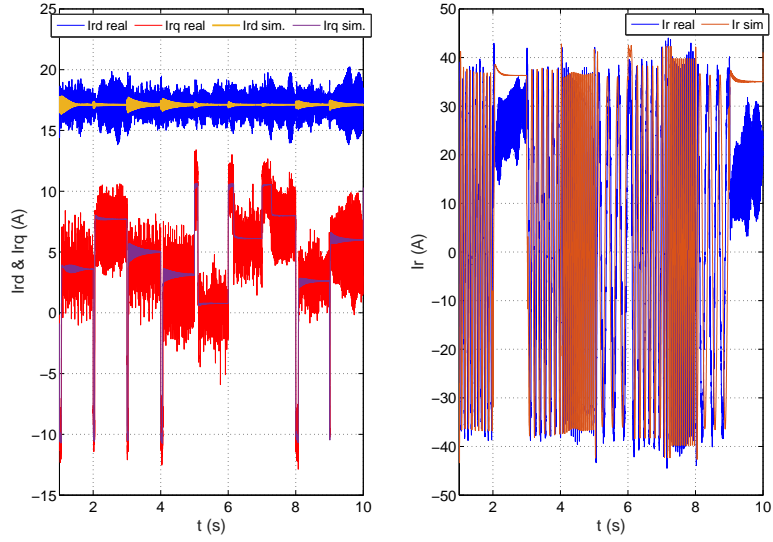


Figure 6: Real and simulated DFIG rotor d and q current components and rotor phase current for the speed and torque profile of figure 4

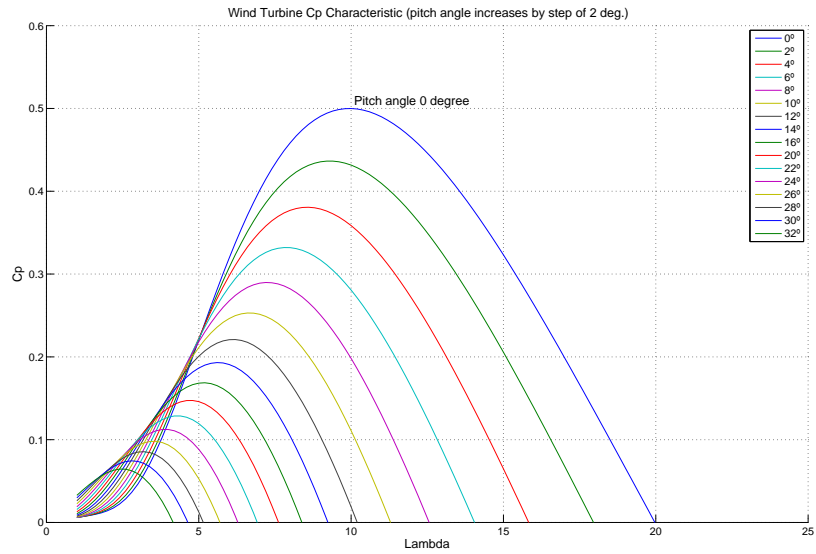


Figure 7: Wind Turbine characteristic curves for the Power Coefficient.

323 platform are compared with the estimated values. Both the wind turbine torque and
 324 speed were estimated with the sliding observer proposed in equation (12). The observer
 325 parameters h_1 and h_2 were firstly tuned using the simulation model and finally refined
 326 at the real platform, obtaining the following values $h_1 = 6200$ and $h_2 = 9000$.

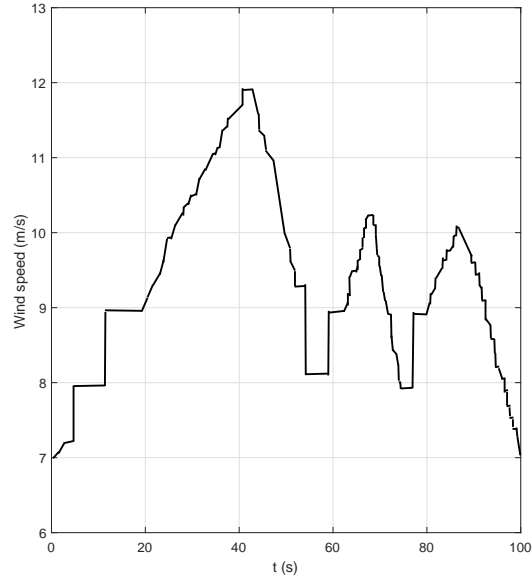


Figure 8: Wind speed profile.

327 Figure 8 presents the random wind speed profile selected for this experimental test.
 328 It should be noted that, although unusual, some sudden changes in the wind speed were
 329 added to analyse the dynamic performance of the observer proposed.

330 Figure 9 shows the DFIG real and estimated rotor speeds as well as the wind turbine
 331 torques obtained for the wind profile represented at figure 8. Comparing both the real
 332 and estimated values, it is clear that the sliding observer estimates accurately the DFIG
 333 torque and speed even when sudden wind speed variations appear, producing sudden
 334 changes at the wind turbine. The estimated torque can be used to impose the optimal
 335 speed reference to the DFIG using equation (26) and, accordingly, to maximize the
 336 power obtained.

337 The moment of inertia almost does not change in a wind system. However, an error
 338 in the total inertia of the system has a relevant impact in the estimation of the torque
 339 in conventional estimators, like the Luenberger observer. The sliding mode observer
 340 provides some robustness under this kind of uncertainties. Figure 10 presents the re-
 341 sponse of the observer to different wind steps changes with mismatches in the inertia
 342 value of 25%. The upper graphs of figure 10 picture the reference calculated for the
 343 optimal speed under different wind steps changes. The line marked as Ref. represents

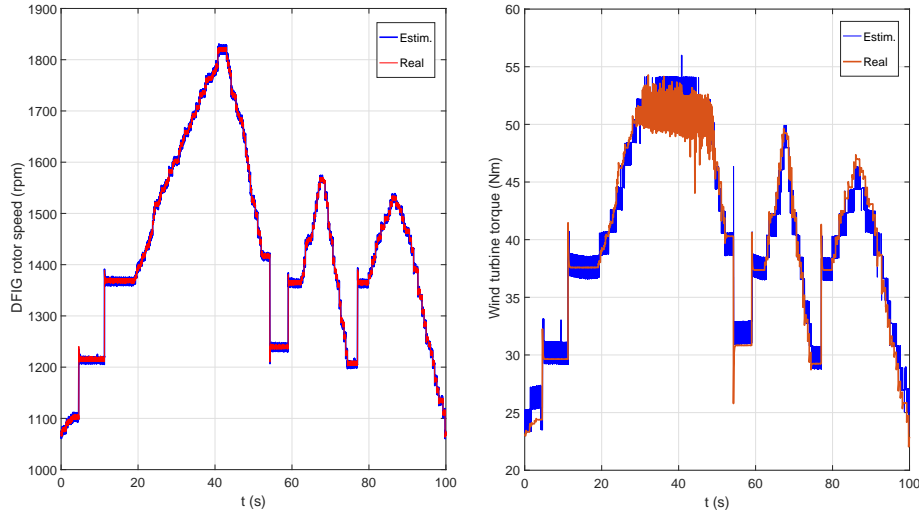


Figure 9: Real and estimated rotor speed and real and estimated wind turbine torque.

344 the optimal speed reference obtained from equation (2), using the wind speed and the
 345 λ_{opt} value. The other lines depict the optimal speed obtained by means of the estimated
 346 torque value, calculated from equations (12) and (26). These figures show the response
 347 of the observer for different error values of the moment of inertia: without error (blue
 348 line), +25% error (red line) and -25% error (green line). It can be appreciated that in
 349 case of error at the moment of inertia, the estimated torque and the optimal speed val-
 350 ues degrade in the transitory for a some time. However, after a short period of time,
 351 the proposed observer behaves correctly yielding the optimal speed reference for the
 352 DFIG. The bottom graphs of this figure show the real torque produced by the wind and
 353 the estimated torque value obtained from the proposed sliding mode observer.

354 Figure 11 shows a comparison between the wind torque value, calculated from
 355 the wind turbine model with eqn.(3), and the wind torque value, obtained using the
 356 proposed sliding mode observer, eqn.(12), when a random wind speed profile is used as
 357 input. Both values are quite similar, hence, we can assume that the proposed observer
 358 provides a good estimation of the wind torque value. The estimated torque allows
 359 to calculate the optimal speed value, with eqn.(26), which can be compared with the
 360 optimal speed value obtained from eqn.(2). This figure also shows that the optimal
 361 speed value obtained from the proposed estimator is good enough. Finally, this figure
 362 shows the behaviour of the system when the pitch angle value changes, which avoids
 363 damages in the system caused when the power generated from the wind exceeds the
 364 DFIG maximum power. Even with these variations at the pitch angle the optimal speed
 365 value is correctly calculated.

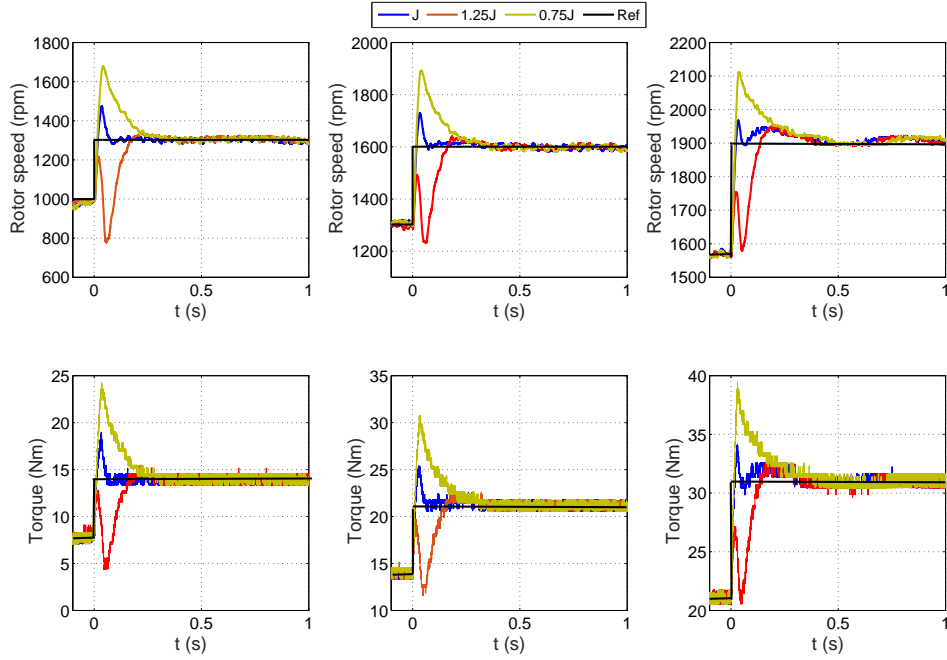


Figure 10: Proposed SMC observer performance for different wind steps changes and under different errors in the inertia moment value

366 5.4. Sliding Mode Control

367 This subsection describes the tests carried out over the experimental platform using
 368 the proposed sliding mode control and torque observer to set the optimal turbine speed.
 369 A scenario in which the wind speed varies randomly, as seen in the figure 12-a, has
 370 been selected. The chosen values for the sliding mode controller were $k = 20$ and
 371 $\beta = 100$. A fixed reactive power of 4000 VAR was used at the DFIG.

372 Figure 12-b compares the values of the wind torque, experimentally measured, and
 373 the sliding mode observer torque, calculated with eqn. (12). The figure shows that the
 374 proposed observer yields a good estimation of the torque value. Figure 12-c shows the
 375 optimal rotor speed obtained from equation (26) using the estimated torque value. This
 376 value is compared with the real optimal reference, obtained from equation (2) using
 377 the real wind speed. As it can be observed, the optimal rotor speed obtained by our
 378 approach is very close to the optimal wind speed value, even though in our case it was
 379 not required. Figure 12-d compares the experimental rotor speed with the reference in
 380 order to evaluate the performance of the proposed sliding mode speed control.

381 Figure 13 shows the stator power, the rotor power and the total power extracted for
 382 the test wind speed profile using the proposed sliding mode controller and observer.

383 Finally, figure 14 shows the stator and rotor currents of the DFIG for the analysed
 384 set up.

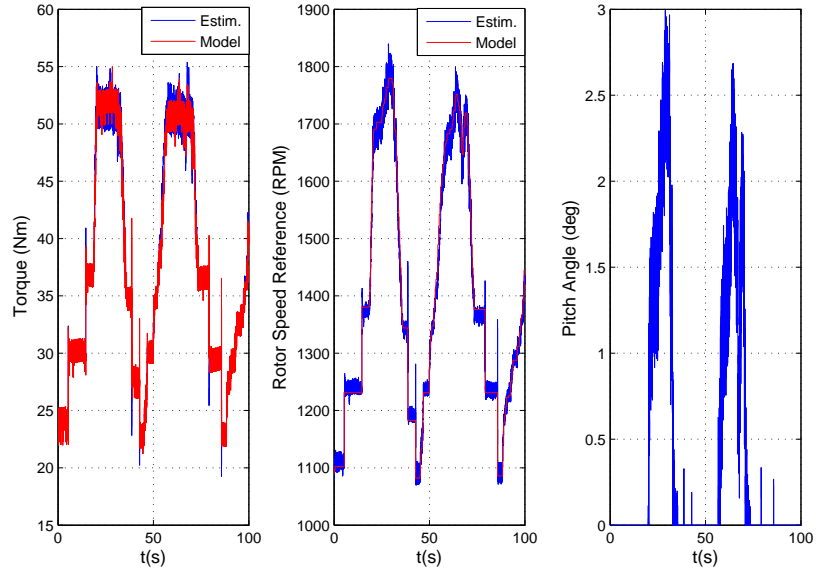


Figure 11: Turbine model torque and SMC estimated torque. Optimal rotor speed according to the turbine model and to the SMC observer. Pitch angle for the imposed torque and speed.

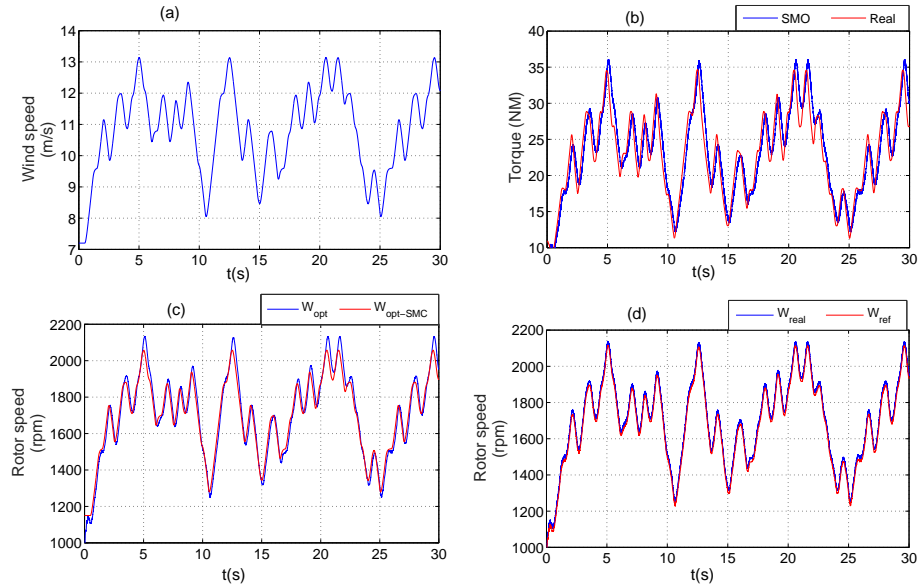


Figure 12: Performance of the SM controller and observer for a random wind profile

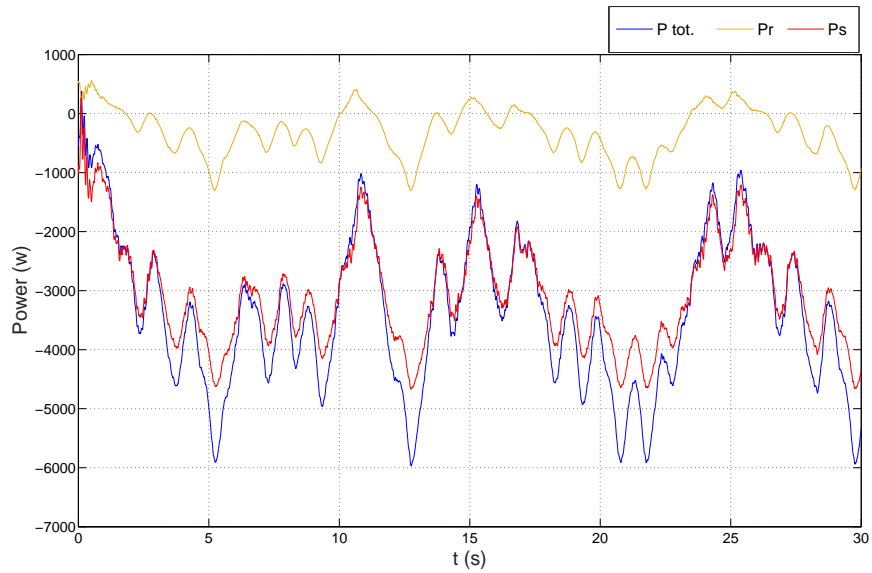


Figure 13: Stator power, the rotor power and the total power for a random wind profile

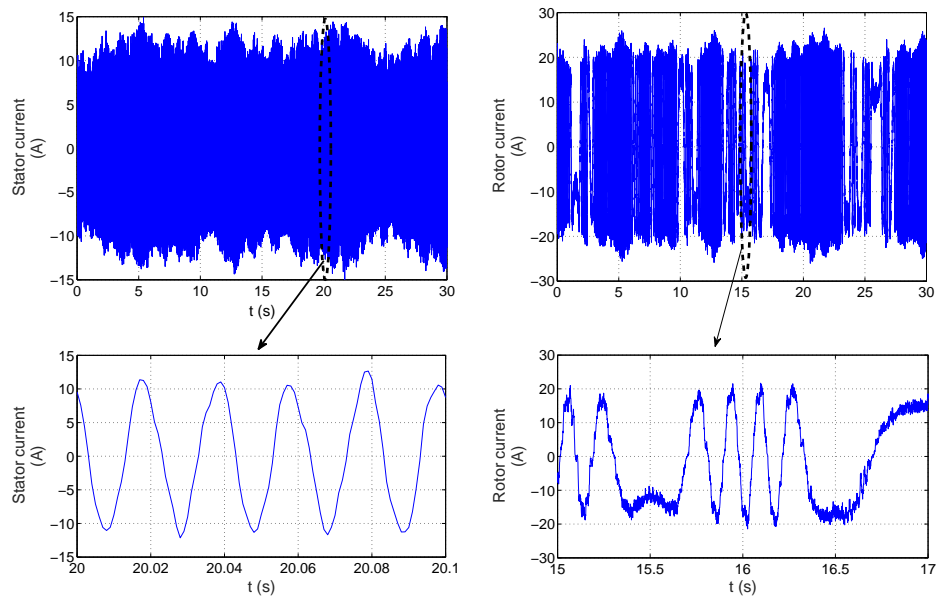


Figure 14: Stator and rotor currents for a random wind profile

385 *5.5. Sliding Mode Control versus PI control*

386 In this experimental validation, the traditional PI controller is compared with the

387 proposed SMC in order to shown the controller performance. As it is well known, the
 388 PI controller is the most widely used controller in the industry and this controller is
 usually employed for comparing the new control schemes [30], [31], [32].

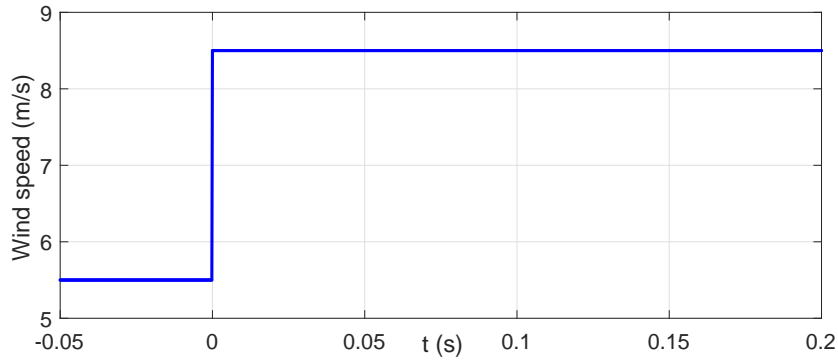


Figure 15: Wind speed step reference

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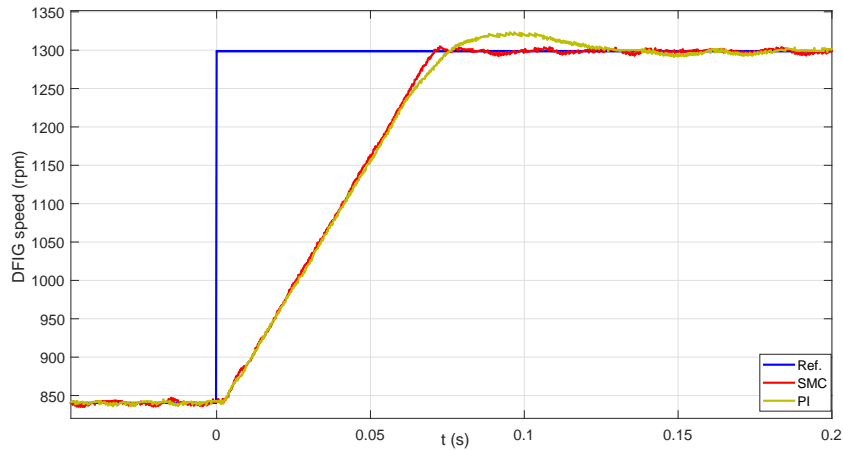


Figure 16: Rotor speed regulation using the traditional PI controller and the proposed SMC

390 This validation has been carried out over the real platform using the step change
 391 in the wind speed, shown in Figure 15, in order to show the controller performance
 392 under a sudden wind changes which is a exigent task for the control scheme. This wind
 393 step variation produces a change in the rotor speed reference (to follow the maximum
 394 power extraction) as it is shown in Figure 16. In the same figure the response for both
 395 controllers is also shown. The PI controller is adjusted with a bandwidth of 50 rad/s and
 396 a phase margin of 75 deg. In this figure it can be observed that using the SMC controller
 397 the rotor speed tracks the reference speed that provides the maximum power extraction

398 from the wind. Obviously due to the system mechanical inertia, the rotor speed can
399 not track the speed changes in the reference speed but after a 0.08 s the reference
400 speed is reached. The figure also shown that the response rate of the PID controller
401 is similar to the response rate of the SMC controller and both controllers provides a
402 similar rise time. However, in the PID controller an overshoot can be observed, and
403 therefore the system takes more time to reach the reference speed, that provides the
404 maximum power extraction, from this new wind speed value. This overshoot can be
405 reduced decreasing the proportional action of the controller but in this case a slower
406 response will be obtained.

407 **6. Conclusion**

408 This paper describes the design of a sliding mode vector control for a doubly feed
409 induction generator drive, used in variable speed wind power generation. An integral
410 sliding surface is proposed to relax the requirements at the acceleration, which is usu-
411 ally required in sliding mode schemes for speed control. The nature of the sliding
412 control technique ensures the robustness of the control scheme under the uncertain-
413 ties that may appear in real systems. The stability of the close loop for the presented
414 scheme was analysed and proved by means of Lyapunov stability theory.

415 Moreover, in order to avoid wind velocity measurements, a sliding mode observer
416 for the aerodynamic torque is proposed. This observer will be used to calculate the
417 reference speed of the wind turbine, providing the optimal tip speed ratio for the sliding
418 mode controller, i.e. which maximizes power extraction.

419 The proposed approach allows to operate a wind turbine over a wide range of values
420 for the wind speed while optimizing power efficiency. This controller regulates the
421 speed of the wind turbine by obtaining the optimal tip speed ratio, and consequently,
422 producing the maximum power.

423 Experimental tests, developed in a real test bench which was designed and con-
424 structed ad hoc, prove that the proposed control method controls efficiently and suc-
425 cessfully the variable speed wind turbine within a range of normal operational condi-
426 tions. These tests evidence that the proposed observer provides a good estimation of
427 the aerodynamic torque under system uncertainties and wind speed variations. Finally,
428 experimental results also show that the speed tracking objective is achieved in order
429 to maintain the maximum power extraction under wind speed variations and system
430 uncertainties.

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